

Research Article

A Fast and Accurate Analytical Method for Parameter Determination of a Photovoltaic System Based on Manufacturer's Data

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Received 3 January 2020; Accepted 24 April 2020; Published 15 May 2020

Academic Editor: Shuhui Li

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Modeling and simulation of a photovoltaic solar system play a significant role in understanding its behavior in various environmental conditions. Utilization of the datasheet information in modeling and simulation of the PV system correlates the experimental data and the theory that instigate the mathematical predictions of an actual system. A single-diode model gives a simple, fast, and straightforward way of depicting the PV system performance. We have developed a new approach of determining the five unknown parameters of a single-diode model using manufacturer's data at three main points: the open circuit point (OCP), short circuit point (SCP), and the maximum power point (MPP) of the IV and PV curves. The ideality factor (A) and the diode saturation current (I_o) are the key unknown parameters that greatly affect the reduplication of the three main points. The purpose of this study is to evaluate the ideality factor using simple calculation procedure starting from its optimal value (A_o) and other values within the proximity of A_o . The optimal value is obtained by assumptions of negligible series resistance (R_s) and very large shunt resistance (R_{sh}). Therefore, the choice of the other ideality factors in the neighborhood of its optimal value gives rise to different values of R_s , R_{sh} , and I_{ph} that are more realistic in an experimental setup. Positive values of R_{sh} and R_s have been iteratively obtained by utilizing data at maximum power point combined with open and short circuit data. The five unknown parameters have been determined in the proximity of A_o and have been used to plot the PV curve with accuracy and precision of less than 0.5% error of maximum power and less than 0.1% error of V_{oc} of manufacturer's data. The proposed method has been implemented using fast, simple, and accurate procedures using GNU Octave programming software to calculate A_o , I_o , R_s , R_{sh} , and I_{ph} and to execute both R_s - R_{sh} and PV characteristic equations of BP3235T, KC200GT, BP-SX 150, and MSX60 PV modules. The reduced steps employed in the algorithm improve execution speed, thereby reducing the computation time.

1. Introduction

Single- and double-diode PV models have been widely used for evaluation and analysis of electrical characteristics for photovoltaic (PV) solar modules [1–3]. Several authors [4–8] have reviewed the advantages and drawbacks of both models and compared their accuracy in replicating the experimental IV and PV curves. Comparative study of the two models showed a trade-off between accuracy and computation time [9]. The double-diode model gives better profiling of the PV modules but requires complex computation procedure in

order to extract its seven unknown parameters [10, 11]. However, improved approaches for calculating unknown parameters using seven [12], six [13], five [14], and four parameters [15] for the double-diode models have also been studied. These methods utilize numerical algorithms such as particle swarm optimization or the Newton-Raphson method that requires large amount of data and several mathematical manipulations to evaluate the unknown parameters. Therefore, the single-diode model remains a method of choice where quick and simplified analyses are required for design of maximum power point tracking (MPPT).

A single-diode model has five unknown parameters (A , I_o , I_{ph} , R_s , and R_{sh}) that must be determined to obtain the best solution of the current-voltage equation [16]. Several approaches have been discussed in the literature for determination of the correct values of these five parameters [17, 18]. The simplest model that reduces the number of unknown parameters to three (A , I_o , and I_{ph}) is a single-exponential diode equivalent circuit without any resistance [19–21]. Ignoring the series and shunt resistances gives inaccurate evaluation of the PV module and fails to account for the loss of power due to electrical contacts, base region, and the front layer [22]. A better approach based on four parameters incorporates a series resistance and omits the shunt resistance [23–25]. The five-parameter model gives the best solutions since it accounts for the effect of power losses due to parasitic series and shunt resistances that are embedded in the solar module as a result of fabrication defects [26, 27].

The five-parameter single-diode model is based on a Shockley diode equivalent circuit with series and parallel resistances and is defined by a transcendental equation that cannot be solved explicitly. Several procedures have been developed for evaluation and determination of the lumped parameters, namely, ideality factor (A), saturation current (I_o), photocurrent (I_{ph}), and shunt (R_{sh}) and series (R_s) resistances, using IV characteristic [8]. Experimental data or information available from manufacturer's datasheet is usually utilized in order to derive the necessary nonlinear equations for extraction of the unknown parameters by considering three crucial points at short circuit current, open circuit voltage, and maximum power point of an IV curve [28]. Experimental procedures are expensive and require more time for setup, data collection, and analysis [29]. Nevertheless, evaluation of the five parameters using manufacturer's data has become more popular since the information about the three critical points is readily available at STC and NOCT for module with IEC standards [30–32]. Different approaches have been proposed for extraction of these unknown parameters depending on which parameter is perturbed first, either R_s or A . Traditionally, the straightforward way of getting the five parameters has been based on five nonlinear equations selected from the three critical points and determining the R_s and R_{sh} from the reciprocal of IV slopes at these points [33, 34]. Computer software has been exploited to solve these equations using Lambert W function [35, 36], nonlinear least square (NLS) [37], and Newton–Raphson algorithms [38, 39]. These techniques are comprehensive methods with good accuracy and fast convergence but their main drawbacks are complex implementation procedures and relatively low calculation speed that require more computing power.

Recent advances in computing technology has led to application of evolutionary computation techniques inspired by biological evolution and chemistry/physics-based, swarming-based, and hybridization-based algorithms for extraction of the unknown parameters for single-, double-, or triple-diode models [40]. These soft-computing techniques are problem solvers using trial and error meta-heuristic or stochastic optimization procedures such as

pattern search (PS) [41], genetic algorithm (GA) [42], simulated annealing (SA) [43, 44], particle swarm optimization [45, 46], artificial fish swarm algorithm (AFSA) [47], fuzzy logic [48], and artificial neural networks (ANNs) [49–51]. These methods have disadvantages, either in terms of sophistication and accuracy or in terms of convergence and speed of execution.

Fast and simplified approaches that calculate all model parameters for the single-diode model of the photovoltaic system directly rely on careful evaluation of the underlying analytical and numerical algorithms [52, 53]. Straightforward analytical methods for describing the parameters of PV models are heavily dependent on nonlinear explicit formulas using I_{sc} , I_{mpp} , V_{mpp} , and V_{oc} data in the PV panel datasheet [54]. The simplest and comprehensive approach assumes arbitrary value of ideality factor and determines other parameters by adjusting the curve at I_{sc} , P_{mpp} , and V_{oc} points [55]. An ideal diode has an ideality factor of unity. However, an appropriate and actual diode for modeling PV system has ideality factor values ranging from one to two [56]. An accurate and fast convergence method of extracting the unknown parameters by first analyzing the ideality factor and setting it as the primary parameter and obtaining the loss resistance through a repetitive process was reported in [57]. The authors used a protracted method to arrive at a control signal that was based on the series resistance value. In this paper, we present a novel analytical method for determining the single-diode model parameters directly from the manufacturer's datasheet using explicit equations.

2. A Single-Diode Equivalent Circuit

A single-diode equivalent circuit shown in Figure 1 can be mathematically modeled using equation (1) [16]. A Shockley's diode is connected in parallel to the current source and shunt resistor and in series to R_s and the load.

$$I = I_{ph} + I_o - I_o \exp \frac{q(V + IR_s)}{AN_s kT} - \frac{V}{R_{sh}} - \frac{IR_s}{R_{sh}}, \quad (1)$$

where I is module output current in amperes (A), V is module output voltage in volts (V), q is charge of an electron = $1.602176634 \times 10^{-19} C$, k is Boltzmann's constant = $1.380649 \times 10^{-23} \cdot m^2 s^{-2} \cdot kg K^{-1}$, and T_{STC} is standard temperature = 298.15 K at STC.

3. Mathematical Modeling of a PV System Using Three Critical Points and Their Respective Slopes of an IV Curve

Equation (1) can be evaluated by analyzing the open circuit, short circuit, and the maximum power points whose data are usually available from datasheets.

At short circuit, $I = I_{sc}$, $V = 0$.

In Figure 1, a short circuit without the load gives zero voltage and current $I \approx I_{sc}$ that depend on I_{ph} , A , I_o , T , R_s , and R_{sh} [16, 58]. Therefore, we can rewrite equation (1) as

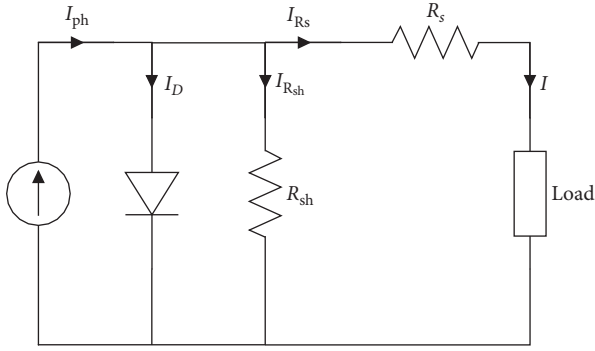


FIGURE 1: A single-diode equivalent circuit.

$$I_{sc} = I_{ph} + I_o - I_o \exp\left(\frac{I_{sc}R_s}{AN_sV_t}\right) - \frac{I_{sc}R_s}{R_{sh}}, \quad (2)$$

where $V_t = kT/q = 0.0256796347$ is the thermal voltage.

At open circuit, $I = 0$, $V = V_{oc}$; likewise, at open circuit, we can also rearrange equation (1) to get

$$I_{ph} = I_o \left(\frac{V_{oc}}{AN_sV_t} \right) - I_o + \frac{V_{oc}}{R_{sh}}. \quad (3)$$

For fast convergence, equation (3) can be expressed logarithmically as

$$\ln\left(I_{ph} + I_o - \frac{V_{oc}}{R_{sh}}\right) - \ln(I_o) = \frac{V_{oc}}{AN_sV_t}. \quad (4)$$

At maximum power point, $I = I_{mpp}$, $V = V_{mpp}$; we can similarly replace $I = I_{mpp}$, $V = V_{mpp}$ in equation (1) to obtain

$$I_{mpp} = I_{ph} + I_o - I_o \exp\left(\frac{V_{mpp} + I_{mpp}R_s}{AN_sV_t}\right) - \frac{V_{mpp} + I_{mpp}R_s}{R_{sh}}. \quad (5)$$

This can also be evaluated as

$$\ln\left(I_{ph} + I_o - I_{mpp} - \frac{V_{mpp} + I_{mpp}R_s}{R_{sh}}\right) - \ln(I_o) = \frac{V_{mpp} + I_{mpp}R_s}{AN_sV_t}. \quad (6)$$

Subtracting equations (4) and (6) yields

$$\ln\left(\frac{I_{ph} + I_o - V_{oc}/R_{sh}}{I_{ph} + I_o - I_{mpp} - (V_{mpp} + I_{mpp}R_s)/R_{sh}}\right) = \frac{V_{oc} - V_{mpp} - I_{mpp}R_s}{AN_sV_t}. \quad (7)$$

This reduces to the five-parameter mathematical model that can be written as

$$A = \frac{V_{oc} - V_{mpp} - I_{mpp}R_s}{N_sV_t \left[\ln\left(\frac{I_{ph} + I_o - V_{oc}/R_{sh}}{I_{ph} + I_o - I_{mpp} - (V_{mpp} + I_{mpp}R_s)/R_{sh}}\right) \right]}. \quad (8)$$

3.1. Photocurrent (I_{ph}) Analysis. The photocurrent can be deduced by rearranging equation (2) in order to obtain equation (9) in terms of A , I_o , R_s , and R_{sh} to get

$$I_{ph} = I_{sc} - I_o + I_o \exp\left(\frac{I_{sc}R_s}{AN_sV_t}\right) + \frac{I_{sc}R_s}{R_{sh}}. \quad (9)$$

The light-generated current of the PV cell is linearly dependent on the solar irradiance and also varies with the temperature. Therefore, the relationship between photocurrent, temperature, and irradiance can be evaluated using temperature coefficient of I_{sc} (K_I) as reported in [38, 59, 60]. Thus,

$$I_{ph} = \frac{s}{s_{STC}} \left[I_{phSTC} + K_I(T - T_{STC}) \right], \quad (10)$$

where T is the actual temperature in Kelvin, T_{STC} is the temperature at STC usually 298.15 K, s is the irradiance on the PV surface, s_{STC} is the irradiance at STC usually 1000 W/m², and I_{phSTC} is the light-generated current at STC.

3.2. Ideality Factor (A) Analysis. One of the simplest ways of determining the ideality factor is to evaluate its value at standard test conditions (STCs) using data provided in manufacturer's datasheet. Using the information obtainable from the datasheet for I_{sc} , I_{mpp} , V_{mpp} , and V_{oc} , the optimal value of ideality factor can be easily deduced from equation

(8) by letting the photocurrent (I_{ph}) to be equivalent to short circuit current (I_{sc}). We can also assume that R_s has minimal resistance value ($R_s \approx 0$) and R_{sh} has large resistance value ($R_{sh} \approx \infty$) whose effects can initially be ignored in the equation. These assumptions will be reevaluated later in the following sections to consider the nominal operation parameters, in which R_s and R_{sh} have other more practical values. Thus, we can evaluate ideality factor as

$$A = \frac{V_{oc} - V_{mpp}}{N_sV_t} \left[\ln\left(\frac{I_{sc} + I_o}{I_{sc} + I_o - I_{mpp}}\right) \right]^{-1}. \quad (11)$$

A closer look at its denominator infers that the logarithm term is greatly influenced by the short circuit current. Therefore, small values of saturation current in microamperes range can be ignored. This gives rise to the optimum value of ideality factor (A_o) that can be expressed as

$$A_o = \frac{V_{oc} - V_{mpp}}{N_sV_t} \left[\ln\left(\frac{I_{sc}}{I_{sc} - I_{mpp}}\right) \right]^{-1}. \quad (12)$$

Figure 2 illustrates the variation of ideality factor (A) with respect to saturation current for BP3235T, KC200GT, BP-SX 150, and MSX60 PV modules, respectively. The values of I_o have been arbitrarily chosen within an acceptable range from 0 to 1 A. According to Figure 2, the ideality factor

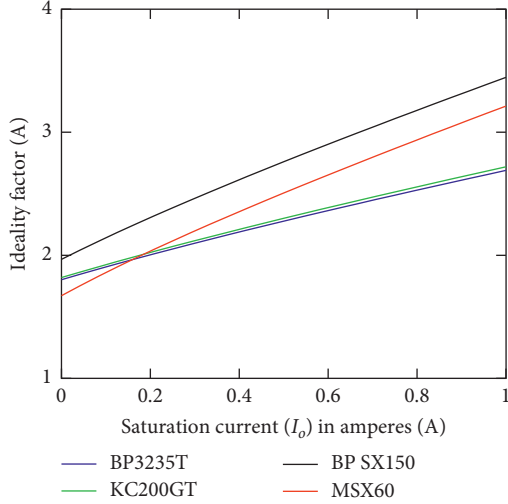


FIGURE 2: A graph of A vs I_o for BP3235T, KC200GT, BP-SX 150, and MSX60 PV modules.

increases as saturation current (I_o) increases and vice versa. The intercept on the y -axis represents the values of optimal ideality factor (A_o) and optimal saturation current ($I_{o\text{opt}}$) for each respective PV modules provided in Table 1.

The optimum value of the ideality factor (A_o) has been deduced from very low series and high shunt resistances. In order to consider other R_s and R_{sh} resistances that are more realistic and achievable in a practical setup, it is important to consider the ideality factor in the neighborhood of A_o such that $1 \leq A \leq A_o$ at a constant I_o . In this work, we have evaluated the effect of ideality factor on I_{ph} , R_s , and R_{sh} by arbitrarily selecting different values of A in the neighborhood of A_o as discussed in Sections 3.3 and 3.4.

$$[I_o]_{I_{sc}, P_{mpp}} = \frac{V_{mpp} + I_{mpp}R_{sh} + I_{mpp}R_s - I_{sc}R_s - I_{sc}R_{sh}}{R_{sh} [\exp(I_{sc}R_s/AN_sV_t) - \exp((V_{mpp} + I_{mpp})R_s/AN_sV_t)]}. \quad (15)$$

Returning to $R_s \approx 0$ and $R_{sh} \approx \infty$, equation (15) reduces to

$$\left[I_{o\text{opt}} \right]_{I_{sc}, P_{mpp}} = \frac{I_{sc} - I_{mpp}}{\exp(V_{mpp}/AN_sV_t)}. \quad (16)$$

$$[I_o]_{V_{oc}, P_{mpp}} = \frac{V_{mpp} - V_{oc} + I_{mpp}R_{sh} + I_{mpp}R_s}{R_{sh} [\exp(V_{oc}/AN_sV_t) - \exp((V_{mpp} + I_{mpp})R_s/AN_sV_t)]}. \quad (17)$$

Again, assuming $R_s \approx 0$ and $R_{sh} \approx \infty$, we can rewrite equation (17) as

$$\left[I_{o\text{opt}} \right]_{V_{oc}, P_{mpp}} = \frac{I_{mpp}}{[\exp(V_{oc}/AN_sV_t) - \exp(V_{mpp}/AN_sV_t)]}. \quad (18)$$

TABLE 1: PV module data (P_{mpp} , I_{sc} , I_{mpp} , V_{oc} , V_{mpp} , and N_s) at STC.

Parameters	PV module			
	Solarex MSX60	BP-SX 150	Kyocera KC200GT	BP3235T
P_{mpp} (W)	60	150	200	235
I_{sc} (A)	3.8	4.75	8.21	8.48
I_{mpp} (A)	3.5	4.35	7.61	7.89
V_{oc} (V)	21.1	43.5	32.9	37.2
V_{mpp} (V)	17.1	34.5	26.3	29.8
N_s	36	72	54	60

3.3. *Saturation Current (I_o) Analysis.* The saturation current appears as a constant at STC in equation (1) and can be calculated by merging two of the three equations that have been derived at the crucial points. Substituting I_{ph} appearing in equation 2 with I_{ph} defined in equation 3 as discussed by the authors in [52, 61], the saturation current (I_o) at I_{sc} and V_{oc} can be derived as follows:

$$[I_o]_{I_{sc}, V_{oc}} = \frac{I_{sc}R_{sh} + I_{sc}R_s - V_{oc}}{R_{sh} [\exp(V_{oc}/AN_sV_t) - \exp(I_{sc}R_s/AN_sV_t)]}. \quad (13)$$

Since I_o depends on temperature, we can first determine its optimal value at STC by assuming the boundary parameter values for $R_s \approx 0$ and $R_{sh} \approx \infty$. Consequently, we can rewrite equation (13) as

$$\left[I_{o\text{opt}} \right]_{I_{sc}, V_{oc}} = \frac{I_{sc}}{[\exp(V_{oc}/AN_sV_t)]}. \quad (14)$$

Similarly, equation (2) can be merged with equation (5) at I_{sc} and P_{mpp} to obtain

Finally, considering equation (3) and (5) at V_{oc} and P_{mpp} , we can deduce the saturation current as

3.3.1. *Dependence of the Saturation Current on Irradiance and Temperature.* The saturation current is not explicitly dependent on the irradiance, so in the absence of irradiance, the term dark saturation current is used interchangeably with saturation current [62, 63]. In contrast, the dark saturation current strongly depends on temperature, geo-

metrical or transport semiconductor parameters, and the intrinsic carrier concentration [63, 64]. In addition, the intrinsic carrier concentration number depends on the conduction and valence band density of states and on the energy band gap (E_g) of the semiconductor [64].

Therefore, saturation current density for a Schottky junction diode can be expressed using the derivation given in [63] as

$$J_o = DT^3 \exp\left(\frac{-qE_g}{kT}\right), \quad (19)$$

where J_o is the dark saturation current density, T is the actual temperature, D is the diode diffusion factor which is independent of temperature, k is Boltzmann's constant, and E_g is the semiconductor's band gap energy ($E_g = 1.12$ eV for Si and 1.35 eV for GaGs).

Applying equation (19) to a solar system, the saturation current can be determined using similar approach presented in [65] to get

$$I_o = I_{o_{STC}} \left[\frac{T}{T_{STC}} \right]^3 \exp \frac{qE_g}{AN_s k} \left[\frac{1}{T_{STC}} - \frac{1}{T} \right], \quad (20)$$

where $I_{o_{STC}}$ is the saturation current at standard test condition.

The saturation current at STC ($I_{o_{STC}}$) can be calculated using either equation (13), (15), or (17). However, these equations depend on R_s , R_{sh} , and A , which are also unknown parameters that must be determined. Thus, since this new approach takes into account the upper boundary value of the ideality factor (A_o) by assuming minimal series resistance ($R_s \approx 0$) and large shunt resistance ($R_{sh} \approx \infty$), same assumptions can be used to calculate the optimal saturation current ($I_{o_{opt}}$) at STC using either equation (14), (16), or (18). Therefore, equation (20) can be written as

$$I_o = I_{o_{optSTC}} \left[\frac{T}{T_{STC}} \right]^3 \exp \frac{qE_g}{AN_s k} \left[\frac{1}{T_{STC}} - \frac{1}{T} \right]. \quad (21)$$

$$R_{sh} = \frac{V_{mpp} + I_{mpp}R_s - I_{sc}R_s}{I_{sc} - I_{mpp} - I_o \exp\left(\frac{V_{mpp} + I_{mpp}R_s}{AN_s V_t}\right) + I_o \exp\left(\frac{I_{sc}R_s}{AN_s V_t}\right)}. \quad (24)$$

Equations (22) and (24) give similar positive R_s and R_{sh} values for ideality factors above A_o since they depend on P_{mpp} and I_{sc} . Equation (23) has unreasonable outcome of R_{sh} and R_s pairs.

(iv) Combining the equations defining open circuit point at (V_{oc}) and maximum power point at (P_{mpp}):

3.4. Series Resistance (R_s) and Shunt Resistance (R_{sh}) Analysis. The shunt and series resistances have been neglected in calculating the optimal ideality factor in the previous section. However, the maximum power of a solar module can be estimated as the optimal power in the absence of R_s and R_{sh} and the power losses in them. The shunt and series can be evaluated using the combinations of equations derived using the three crucial points and the slopes at either output power at maximum power point or at short circuit point. The following section discusses these different ways of deriving the relationship between R_{sh} and R_s .

(i) At maximum power point only: the relationship between R_{sh} and R_s can be evaluated by rearranging equation (5) to get

$$R_{sh} = \frac{V_{mpp} + I_{mpp}R_s}{I_{ph} - I_{mpp} - I_o \left(\exp\left(\frac{V_{mpp} + I_{mpp}R_s}{AN_s V_t}\right) - 1 \right)}. \quad (22)$$

(ii) Combining the equations defining the short circuit point at (I_{sc}) and open circuit point at (V_{oc}): equating (3) and (4) correlates R_{sh} and R_s with I_{sc} and V_{oc} as follows:

$$R_{sh} = \frac{V_{oc} - I_{sc}R_s}{I_{sc} + I_o \exp\left(\frac{I_{sc}R_s}{AN_s V_t}\right) - I_o \exp\left(\frac{V_{oc}}{AN_s V_t}\right)}. \quad (23)$$

(iii) Combining the equations defining short circuit point at (I_{sc}) and maximum power point at (P_{mpp}): this can be done by substituting equation (3) into equation (5) to obtain

The dependence of R_{sh} on R_s , V_{oc} and P_{mpp} can be expressed by rearranging equations (4) and (5) to obtain

$$R_{sh} = \frac{V_{oc} - V_{mpp} - I_{mpp}R_s}{I_{mpp} + I_o \exp\left(\frac{V_{mpp} + I_{mpp}R_s}{AN_s V_t}\right) - \exp\left(\frac{V_{oc}}{AN_s V_t}\right)}. \quad (25)$$

- (v) Using the derivatives of equation (1) with respect to V at short circuit and open circuit points along with the derivative of $P=IV$ with respect to V or I at maximum power point [16, 38, 58, 66]:

Differentiating equation (1) with respect to V gives

$$\frac{\partial I}{\partial V} = -\frac{I_o}{AN_s V_t} \left\{ \left(1 + \frac{\partial I}{\partial V} R_s \right) \exp\left(\frac{V + IR_s}{AN_s V_t}\right) - \frac{1}{R_{sh}} \left(1 + \frac{\partial I}{\partial V} R_s \right) \right\}. \quad (26)$$

It has been shown in [67] that the derivative at short circuit point can be approximated as

$$\left[\frac{\partial I}{\partial V} \right]_{I=I_{sc}} = -\frac{1}{R_{sh}}, \quad (27)$$

and at open circuit point, the derivative can be approximated as

$$\left[\frac{\partial I}{\partial V} \right]_{V=V_{oc}} = -\frac{1}{R_s}. \quad (28)$$

At maximum power point, the power derivative with respect to voltage can be evaluated as

$$\frac{\partial P}{\partial V} = \left(\frac{\partial I}{\partial V} \right) V + I = 0. \quad (29)$$

Taking into consideration that $V=V_{mpp}$ and $I=I_{mpp}$ at maximum power point, we can rewrite equation (29) as

$$\begin{aligned} \frac{I_{mpp}}{V_{mpp}} = & -\frac{I_o}{AN_s V_t} \left\{ \left(1 - \frac{I_{mpp}}{V_{mpp}} R_s \right) \exp\left(\frac{V_{mpp} + I_{mpp} R_s}{AN_s V_t}\right) \right\} \\ & - \frac{1}{R_{sh}} \left(1 - \frac{I_{mpp}}{V_{mpp}} R_s \right). \end{aligned} \quad (30)$$

Rearranging equation (30), we can obtain

$$R_{sh} = \frac{V_{mpp} - I_{mpp} R_s}{I_{mpp} - I_o / AN_s V_t (V_{mpp} - I_{mpp} R_s) \exp\left(\frac{V_{mpp} + I_{mpp} R_s}{AN_s V_t}\right)}. \quad (31)$$

The four independent equations (22), (24), (25), and (31) that depend on V_{oc} , I_{mpp} , and V_{mpp} , have four unknown parameters (R_s , R_{sh} , A , and I_o) that can be resolved using either Newton–Raphson algorithm [38, 39, 68, 69], non-linear least square (NLS) algorithm [37], or Lambert W function [35, 36, 70] algorithm to analytically determine the values of unknown PV model parameters. These methods have different levels of computational complexity, accuracy, convergence, and execution speed [8]. A simplified approach is presented in this work using an iterative algorithm that considers the boundary values of ideality factor and saturation current as discussed in the following section.

3.4.1. Evaluation of R_{sh} and R_s Based on A below A_o (Method 1). This paper proposes a robust method of obtaining the unknown parameters of equation (1) by first obtaining the optimal values of the ideality factor using equation (12) and consequently obtaining $I_o \approx I_{o_{opt}}$ based on either equation (14), (16), or (18). The actual ideality factor (A) is arbitrarily

chosen in the neighborhood of A_o and applied in either equation (25) or (31), which leaves R_s and R_{sh} as the only unknown parameters. Utilization of equations (25) and (31) can be done separately when the values of A are selected above or below (A_o). When applied in equation (25), the values of ideality factor below A_o give positive and negative values of R_{sh} but give negative and unrealistic values for A_s above A_o . The best choice of R_s and R_{sh} pairs should be at positive peak values of R_{sh} . A choice of R_s values in steps of $+0.0001\Omega$ above R_s of a positive peak gives R_s and R_{sh} pairs that are more pragmatic. This approach is a departure from the conventional analytical methods reported in [52, 55, 58, 61] that assumes arbitrary values of ideality factor between 1 and 1.5. The optimal saturation current is kept constant when arbitrarily selecting the ideality factor near A_o . To ensure a fast and an easy way of evaluating R_s and R_{sh} , I_o in equation (25) can be replaced by $I_{o_{opt}}$ of equation (14) to get

$$R_{sh} = \frac{V_{oc} - V_{mpp} - I_{mpp} R_s}{I_{mpp} + I_{sc} \exp(-V_{oc}/A_o N_s V_t) \exp\left(\frac{V_{mpp} + I_{mpp} R_s}{AN_s V_t}\right) - I_{sc} \exp(-V_{oc}/A_o N_s V_t) \exp(V_{oc}/AN_s V_t)}. \quad (32)$$

Equation (32) can be analyzed using a similar approach used in [55], in which an iterative method is employed to obtain the $[R_s, R_{sh}]$ pair that satisfies equation (33) in which

P_{mpp} relationship has been derived by replacing $I=I_{mpp}$ and $V=V_{mpp}$ in (1) to yield

$$P_{\text{mpp}} = I_{\text{mpp}} V_{\text{mpp}} = V_{\text{mpp}} \left(I_{\text{ph}} + I_o - I_o \exp \left(\frac{V_{\text{mpp}} + I_{\text{mpp}} R_s}{A N_s V_t} \right) - \frac{V_{\text{mpp}} + I_{\text{mpp}} R_s}{R_{\text{sh}}} \right). \quad (33)$$

Unlike in ref. [55] where R_s is arbitrarily selected, this iterative process involves determining the R_s and R_{sh} pair using equation (32) by selecting the values of A that are less than A_o and subsequently obtaining P_{mpp} without necessarily plotting the PV curve. Moreover, the ideality factor is adjusted in order to get an R_s and R_{sh} pair that guarantees $P_{\text{mpp}} = V_{\text{mpp}} I_{\text{mpp}}$ that matches P_{mpp} in the datasheet. This process is repeated until the value of simulated P_{mpp} matches

$$R_{\text{sh}} = \frac{V_{\text{mpp}} - I_{\text{mpp}} R_s}{I_{\text{mpp}} - I_{\text{sc}} / A N_s V_t \exp(-V_{\text{oc}} / A_o N_s V_t) (V_{\text{mpp}} - I_{\text{mpp}} R_s) \exp((V_{\text{mpp}} + I_{\text{mpp}} R_s) / A N_s V_t)}. \quad (35)$$

Equation (35) gives realistic and viable $[R_s, R_{\text{sh}}]$ pairs if and only if the values of ideality factor are above A_o .

3.4.3. Characterization of R_{sh} and R_s . The simplest and most convenient way of displaying the relationship between R_{sh} and R_s is through a graphical method using R_{sh} against R_s plots of equation (32) as illustrated in Figure 3 for BP3235T, Kyocera KC200GT, BP-SX 150, and Solarex MSX60 solar modules.

The values of R_s have been chosen between 0 and 10 ohms to obtain a positive value of R_{sh} using simple code in GNU Octave software. For any choice of ideality factor, there are R_s and R_{sh} pairs with nonzero value of R_s . Further, as illustrated in Figure 3, R_{sh} has positive and negative values for various R_s .

4. Parameter Calculation Procedure

The flowchart shown in Figure 4 gives step by step procedures that have been used to determine all the desired parameters. The immediate action is to obtain the I_{sc} , I_{mpp} , V_{mpp} , and V_{oc} from the datasheet followed by direct calculation of the optimal ideality factor (A_o) using equation (12) and $I_{o_{\text{opt}}}$ using equation (14). R_s and R_{sh} are evaluated using equation (32) for different values of A_s in the neighborhood of A_o such that $1 \leq A \leq A_o$. Consequently, I_{ph} is calculated using equation (9). The actual saturation current (I_o) corresponding to each ideality factor is recalculated using equation (14).

The maximum power (P_{mpp}) is calculated using equation (33) using a simple code in GNU Octave software. Further, V_{oc} error is calculated by comparing the simulated V_{oc} of equation (3) values with the datasheet values. These procedures are repeated several times as shown in Figure 4. However, the process is stopped if a maximum power error is less than 0.5% and/or the short circuit voltage is greater than 0.1%. Calculated values of I_o , R_{sh} , R_s , and I_{ph} and

the maximum power value in the datasheet or has an error margin of less than 0.5% [22].

The value of I_{ph} in equation (33) can be calculated using equation (9). The percentage error in power can be determined by calculating the simulated maximum power minus maximum power using I_{mpp} and V_{mpp} from datasheet as expressed in the following equation:

$$P_{\text{mpp}} \text{Error} = \Delta P_{\text{mpp}} \% = \frac{P_{\text{mpp}} - I_{\text{mpp}} V_{\text{mpp}}}{I_{\text{mpp}} V_{\text{mpp}}} \times 100\%. \quad (34)$$

3.4.2. Evaluation of R_{sh} and R_s Based on A above A_o (Method 2). Similar derivation has been used in equation (31) by replacing I_o with $I_{o_{\text{opt}}}$ of equation (14) to obtain

simulated values of V_{oc} and P_{mpp} and errors for different ideality factors near A_o are summarized in Tables 2–5. Eminently, the instantaneous value of R_s increases as the ideality factor (A_s) decreases. Thus, the most suitable ideality factor should be chosen near A_o provided the errors in P_{mpp} and V_{oc} are within the acceptable limits of less than 0.5% and 0.1%, respectively [22]. The same procedure is repeated when implementing the second approach. The only changes effected are replacement of equation (32) with (35), and ideality factors (A_s) are selected above A_o .

5. Analysis of Extracted Parameters and Simulated Data

Table 2 gives a summary of extracted parameters and simulated data for P_{mpp} and V_{oc} and their respective errors for BP3235T PV solar module. As depicted in equations (32) and (33) and verified in Figure 3, the shunt resistance varies greatly within a small range of series resistance. Henceforth, the series resistance has been used as the lead parameter that controls the interactive process that stops when appropriate R_s is reached which guarantees simulated data match the experimental data. In this work, the experimental data are obtained from the information available in the manufacturer's datasheet.

For BP3235T PV solar module, the value of ideality factor has been kept constant at $A = 1.792$ since it provided satisfactory $[R_s, R_{\text{sh}}]$ pair that resulted to a small error at both maximum power point and V_{oc} . $R_{\text{s3}} = 0.242 \Omega$ presented the least ΔV_{oc} and ΔP_{mpp} errors of 0% and 0.001%, respectively. The values of R_s have been varied and checked within $\pm 1 \text{ m}\Omega$, which are presented in Table 2 as R_{s1} to R_{s4} , and offer satisfactory results as compared to other data published in [25, 71] as discussed later in Section 7.

Tables 3–5 give similar summary for KC200GT, BP-SX 150, and MSX60 PV modules. For KC200GT, BP-SX 150, and MSX60 PV, the values corresponding to $R_{\text{s4}} = 0.2188 \Omega$,

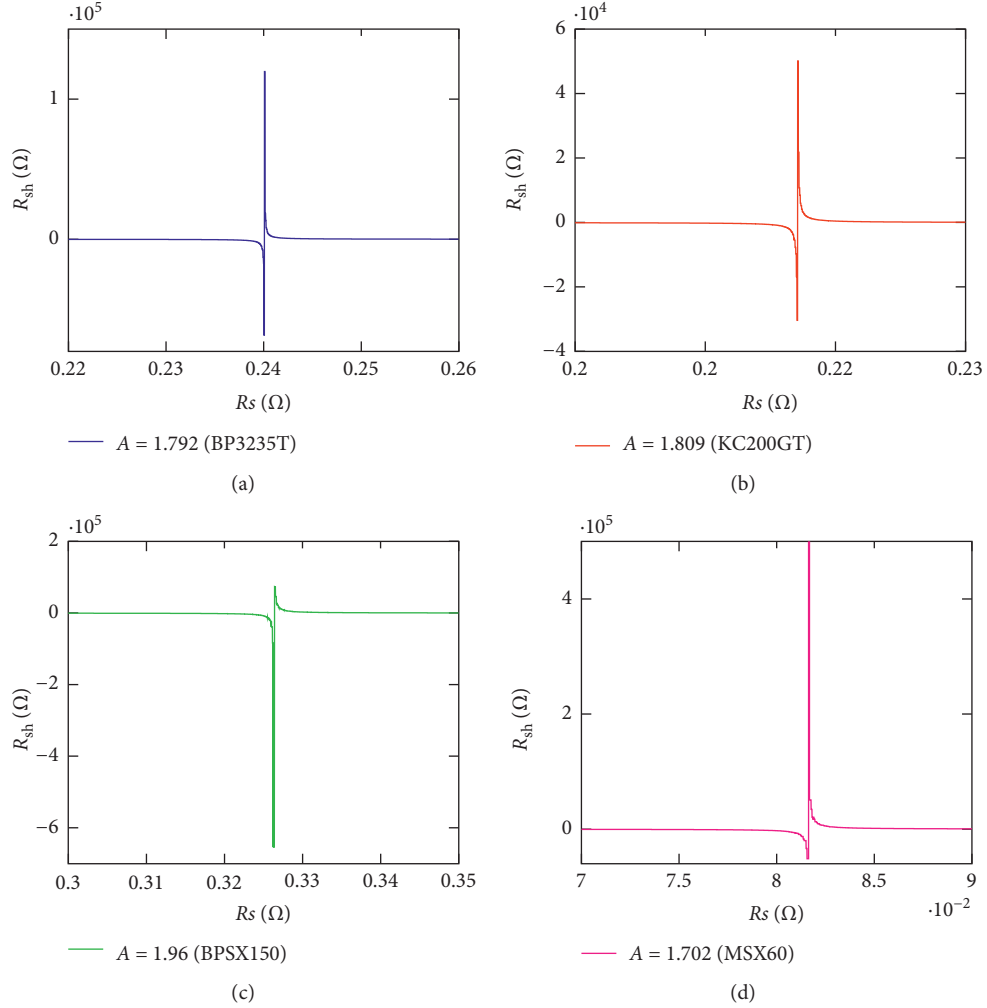


FIGURE 3: A graph of R_{sh} vs R_s at STC for (a) BP3235T, (b) Kyocera KC200GT, (c) BP-SX 150, and (d) Solarex MSX60.

$R_{s4} = 0.3301\Omega$, and $R_{s5} = 0.15\Omega$, respectively, present the least errors of less than 0.1% based on the first method. These modules have been chosen to ensure that the tests cover a wide variety of systems with different output power. This shows how valid the proposed approach is for various PV systems.

In the second procedure, where ideality factors (A_s) are selected above A_o , the same R_s values of $R_{s4} = 0.2188\Omega$, $R_{s4} = 0.3301\Omega$, and $R_{s5} = 0.15\Omega$ have been set as the target for determining the $[R_s, R_{sh}]$ pair. As illustrated in Tables 3–5, the second approach, unlike the first method, results in unsatisfactory data with ΔV_{oc} and ΔP_{mpp} errors reaching and exceeding 0.5 percent. This makes the first approach superior over the second one.

Table 6 gives a summary of appropriate parameters for MSX60, BP-SX 150, KC200GT, and BP3235T.

6. PV Characteristics at Standard Test Conditions

This proposed technique is a straightforward way of evaluating the desired unknown parameters of a single-

diode PV model. The extracted parameters presented in this work can be characterized using IV and/or PV plots. The output power of the solar cell/module can be calculated and evaluated using equation (1) multiplied by the output voltage to obtain

$$P = VI = V \left(I_{ph} + I_o - I_o \exp \left(\frac{V + IR_s}{AN_s V_t} \right) - \frac{V}{R_{sh}} - \frac{IR_s}{R_{sh}} \right). \quad (36)$$

For simplified and clear description of the distinctive variation of PV output power, we have chosen to verify its behavior with respect to identified parameter through PV plots. The PV characteristics are plotted as shown in Figures 5–8 by using the extracted parameters in equation (36) and using a powerful tool/software with built-in plotting and visualization capability such as MATLAB and Octave.

As shown in Figures 5(a), 6(a), 7(a), and 8(a) for ideality factors below A_o , the output power plots are closer to the expected result, closer to the datasheet values. This is in conformity with the data presented in Tables 2–5.

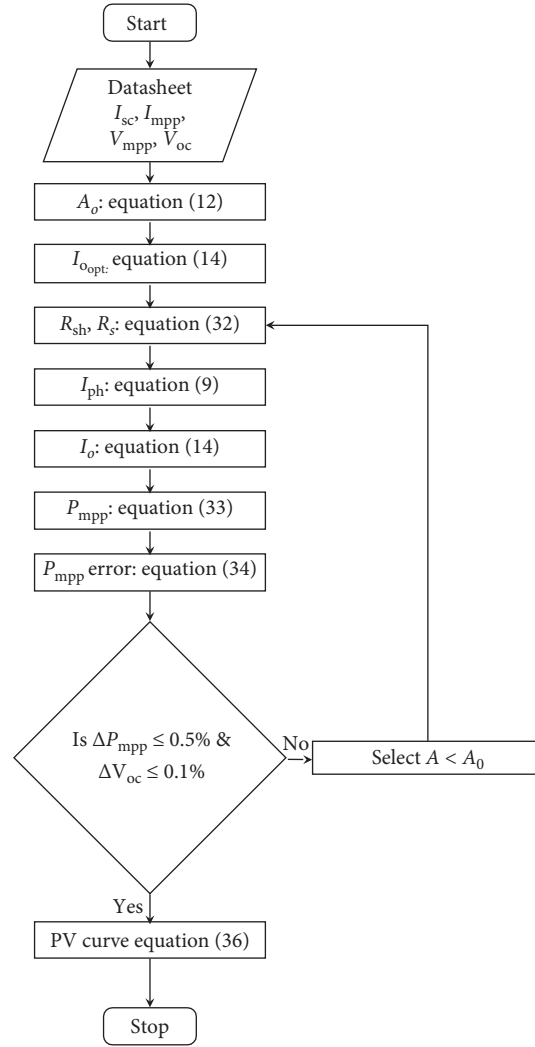
FIGURE 4: An algorithm for evaluating the A_o , A , $I_{o_{opt}}$, I_o , R_{sh} , and R_s using manufacturer's datasheet.

TABLE 2: Extracted parameters and data from simulations for BP3235T.

	R_s (Ω)	$A < A_o$	I_o (A)	R_{sh} (Ω)	I_{ph} (A)	V_{oc} (V)	ΔV_{oc} (%)	P_{mpp} (W)	ΔP_{mpp} (%)
<i>Method 1</i>									
Rs1	0.002	1.801956 (A_o)	1.286446E-5	2195.78	8.4800	37.194	-0.016	234.17	-0.405
Rs2	0.2419	1.792	1.194159E-5	1146.06	8.4818	37.201	0.003	235.17	0.020
Rs3	0.242	1.792	1.194159E-5	1066.37	8.4819	37.2	0	235.12	-0.001
Rs4	0.2421	1.792	1.194159E-5	997.02	8.4821	37.199	-0.003	235.06	-0.026
Rs4	0.2422	1.792	1.194159E-5	936.10	8.4822	37.198	-0.005	235	-0.052
<i>Method 2</i>									
Rs1	0.0852	1.801956 (A_o)	1.286446E-5	6595.76	8.4801	37.2	0	235.42	0.127
Rs2	0.2419	1.8626	1.989991E-5	992.72	8.4821	37.199	-0.003	232.96	-0.920
Rs3	0.242	1.8626	1.989991E-5	1066.30	8.4819	37.199	-0.003	233.03	-0.890
Rs4	0.2421	1.8626	1.989991E-5	1151.68	8.4818	37.2	0	233.09	-0.864
Rs5	0.2422	1.8626	1.989991E-5	1251.95	8.4817	37.201	0.003	233.14	-0.843

The zoomed parts represent the maximum power points and the open circuit voltages, respectively. The curves demonstrate validity of simulated values where the critical points of open circuit voltage and maximum power point have very minimal and acceptable deviation of less than 0.1%.

The plots for Figures 5(b), 6(b), 7(b), and 8(b) whose ideality factors have been selected above A_o clearly show a significant deviation at maximum power point. This is also shown in Tables 2–5 where the errors in maximum power and short circuit voltage surpass 0.5 percent. The data also show errors above 0.5 percent for high-output power modules like

TABLE 3: Extracted parameters and data from simulations for Kyocera KC200GT.

	R_s (Ω)	$A < A_o$	I_o (A)	R_{sh} (Ω)	I_{ph} (A)	V_{oc} (V)	ΔV_{oc} (%)	P_{mpp} (W)	ΔP_{mpp} (%)
<i>Method 1</i>									
Rs1	0.001	1.81928 (A_o)	$1.7807E-5$	3636.875	8.2100	32.897	-0.0091	200.24	0.048
Rs2	0.2186	1.809	$1.65353E-5$	1219.678	8.2115	32.902	0.0061	200.22	0.038
Rs3	0.2187	1.809	$1.65353E-5$	1116.104	8.2116	32.901	0.0030	200.17	0.013
Rs4	0.2188	1.809	$1.65353E-5$	1028.696	8.2118	32.9	0	200.12	-0.011
Rs5	0.2189	1.809	$1.65353E-5$	953.941	8.2119	32.9	0	200.07	-0.036
<i>Method 2</i>									
Rs1	0.0718	1.81928 (A_o)	$1.78074E-5$	7020.18	8.2101	32.9	0	200.66	0.26
Rs2	0.2186	1.882	$2.75010E-5$	612.69	8.2130	32.893	-0.021	197.84	-1.15
Rs3	0.2187	1.882	$2.75010E-5$	644.29	8.2128	32.894	-0.018	197.89	-1.13
Rs4	0.2188	1.882	$2.75010E-5$	679.35	8.2127	32.894	-0.018	197.97	-1.09
Rs5	0.2189	1.882	$2.75010E-5$	718.45	8.2125	32.895	-0.015	198.06	-1.04

TABLE 4: Extracted parameters and data from simulations for BP-SX 150.

	R_s (Ω)	$A < A_o$	I_o (A)	R_{sh} (Ω)	I_{ph} (A)	V_{oc} (V)	ΔV_{oc} (%)	P_{mpp} (W)	ΔP_{mpp} (%)
<i>Method 1</i>									
Rs1	0.004	1.96721 (A_o)	$3.03832E-5$	94045.01	4.7500	43.5	0	150.18	0.070
Rs2	0.3295	1.96	$2.90755E-5$	5001.02	4.7503	43.509	0.0207	150.16	0.057
Rs3	0.3297	1.96	$2.90755E-5$	4555.72	4.7504	43.508	0.0184	150.14	0.043
Rs4	0.3299	1.96	$2.90755E-5$	4183.07	4.7504	43.507	0.0161	150.12	0.030
Rs5	0.3301	1.96	$2.90755E-5$	3866.64	4.7504	43.507	0.0161	150.09	0.010
<i>Method 2</i>									
Rs1	0.1272	1.96721 (A_o)	$3.03832E-5$	10277.75	4.7501	43.499	-0.0023	150.09	0.0100
Rs2	0.3295	2.007	$3.85129E-5$	7860.56	4.7502	43.510	0.0230	149.38	-0.4631
Rs3	0.3297	2.007	$3.85129E-5$	10013.81	4.7502	43.511	0.025	149.42	-0.4364
Rs4	0.3299	2.007	$3.85129E-5$	13793.17	4.7501	43.512	0.028	149.43	-0.4298
Rs5	0.3301	2.007	$3.85129E-5$	22158.12	4.7501	43.513	0.03	149.47	-0.4031

TABLE 5: Extracted parameters and data from simulations for Solarex MSX60.

	R_s (Ω)	$A < A_o$	I_o (A)	R_{sh} (Ω)	I_{ph} (A)	V_{oc} (V)	ΔV_{oc} (%)	P_{mpp} (W)	ΔP_{mpp} (%)
<i>Method 1</i>									
Rs1	0.001	1.70418 (A_o)	$5.79749E-6$	5989.76	3.8000	21.1	0	59.86	0.0184
Rs2	0.147	1.7	$5.60956E-6$	21759.59	3.8000	21.106	0.0284	59.91	0.0986
Rs3	0.148	1.7	$5.60956E-6$	3127.83	3.8002	21.104	0.0190	59.83	-0.0284
Rs4	0.149	1.7	$5.60956E-6$	1681.73	3.8003	21.102	0.0095	59.76	-0.1487
Rs5	0.150	1.7	$5.60956E-6$	1148.49	3.8005	21.100	0.0000	59.68	-0.2840
<i>Method 2</i>									
Rs1	0.0351	1.70418 (A_o)	$5.79749E-6$	3409.80	3.8000	21.1	0	59.87	0.0251
Rs2	0.147	1.737	$7.46677E-6$	733.8600	3.8008	21.095	-0.0237	59.22	-1.0560
Rs3	0.148	1.737	$7.46677E-6$	1052.9594	3.8005	21.098	-0.0095	59.39	-0.7686
Rs4	0.149	1.737	$7.46677E-6$	1866.4501	3.8003	21.102	0.0095	59.47	-0.6383
Rs5	0.150	1.737	$7.46677E-6$	8273.5534	3.8001	21.106	0.0284	59.58	-0.4478

TABLE 6: A summary of appropriate parameters for MSX60, BP-SX 150, KC200GT, and BP3235T.

	MSX60		BP-SX 150		KC200GT		BP3235T	
	Method 1	Method 2	Method 1	Method 2	Method 1	Method 2	Method 1	Method 2
R_s (Ω)	0.150	0.150	0.3301	0.3301	0.2188	0.2188	0.2420	0.2420
R_{sh} (Ω)	1148.49	8273.5533	3866.64	22158.12	1028.695	679.35	1066.37	1066.30
A	1.7	1.737	1.96	2.007	1.809	1.882	1.7920	1.8626
I_o (A)	$5.6096E-6$	$7.4668E-6$	$2.9076E-5$	$3.8513E-5$	$1.6535E-5$	$2.7501E-5$	$1.1942E-5$	$1.99E-5$
I_{ph} (A)	3.800013	3.8001	4.7504	4.7501	8.2119	8.2127	8.481938	8.4819

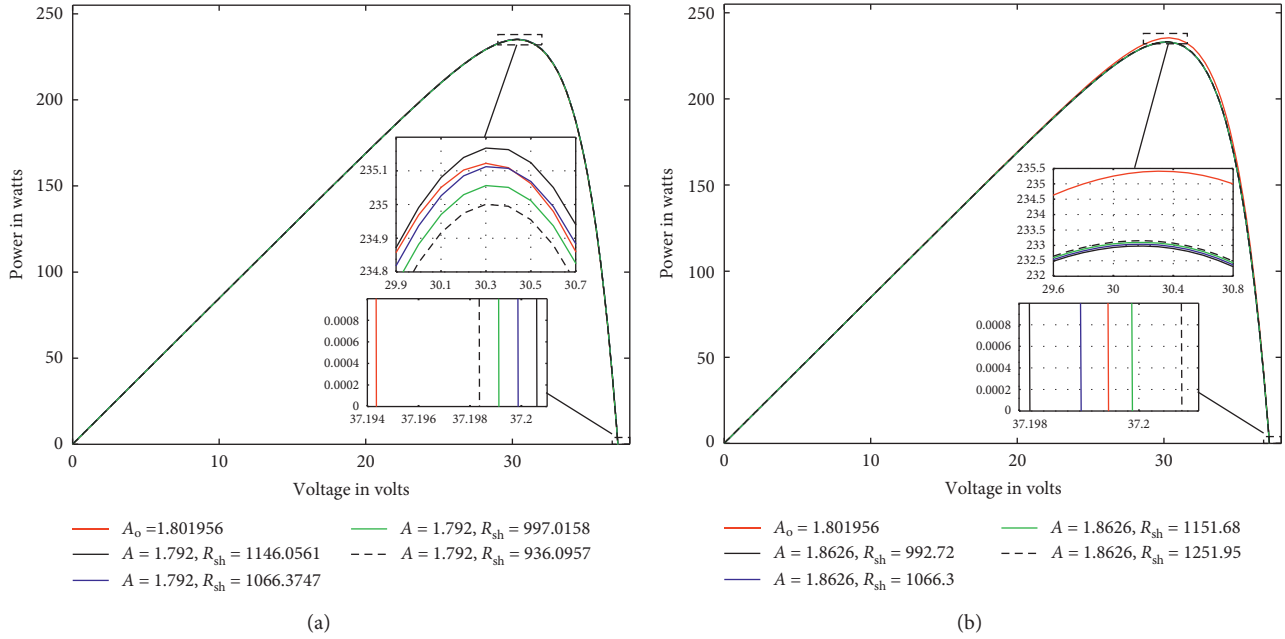


FIGURE 5: PV curves for BP3235T showing different (R_{sh}) values for (a) $A < A_o$ and (b) $A > A_o$.

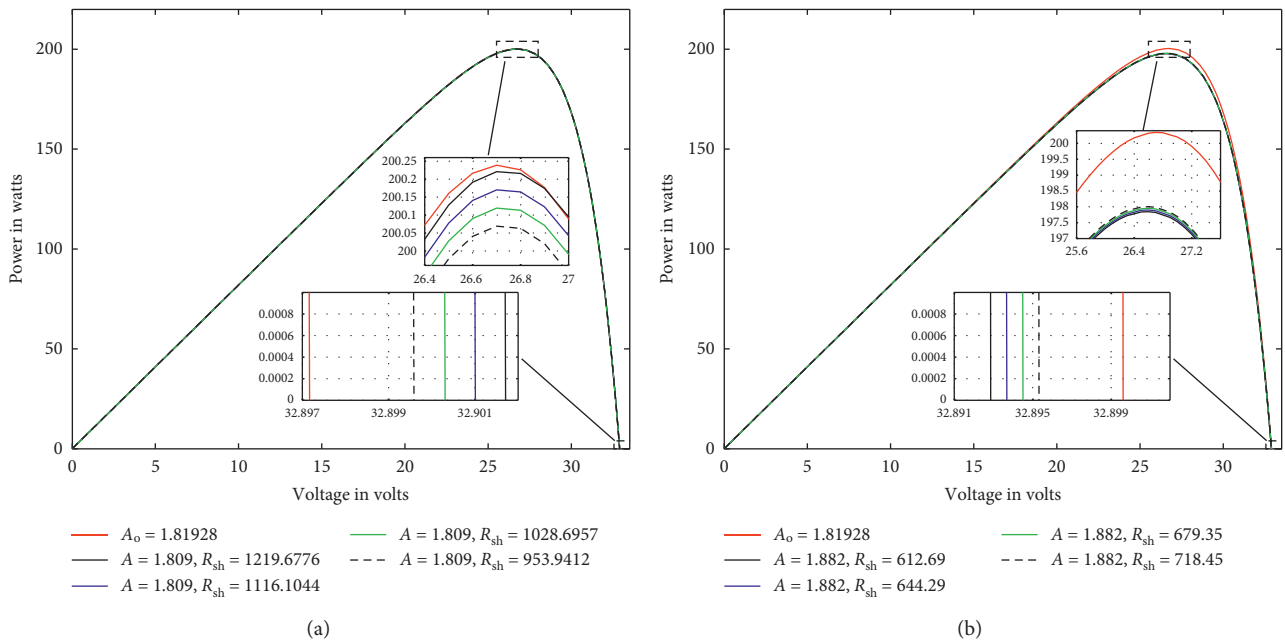


FIGURE 6: PV curves for KC200GT showing different (R_{sh}) values for (a) $A < A_o$ and (b) $A > A_o$.

Kyocera 200GT and BP2235T. In order to overcome this drawback, the ideality factor can be selected very close to A_o and consequently other parameters can be changed which will result in small errors. However, this reduces the values of R_s , and care must be taken not to get nonviable values.

7. Comparison of the Proposed Analytical Method with Other Related Works

In this paper, the ideality factor, the saturation current, and the series resistance are the three key parameters that have

been carefully evaluated and considered to have a major impact on the PV relationship compared to the shunt resistance and I_{ph} . Since the ideality factor has been considered as the immediate parameter to be calculated, the next critically analyzed parameter is the saturation current using equations (13) to (18). Furthermore, due to the nonlinearity of the equations that relate to these parameters, it has been important to use iterative solutions for series and shunt resistances.

As shown in Tables 7–9, ideality factors between 1 and 1.5 have lower I_o values than ideal factors closer to A_o .

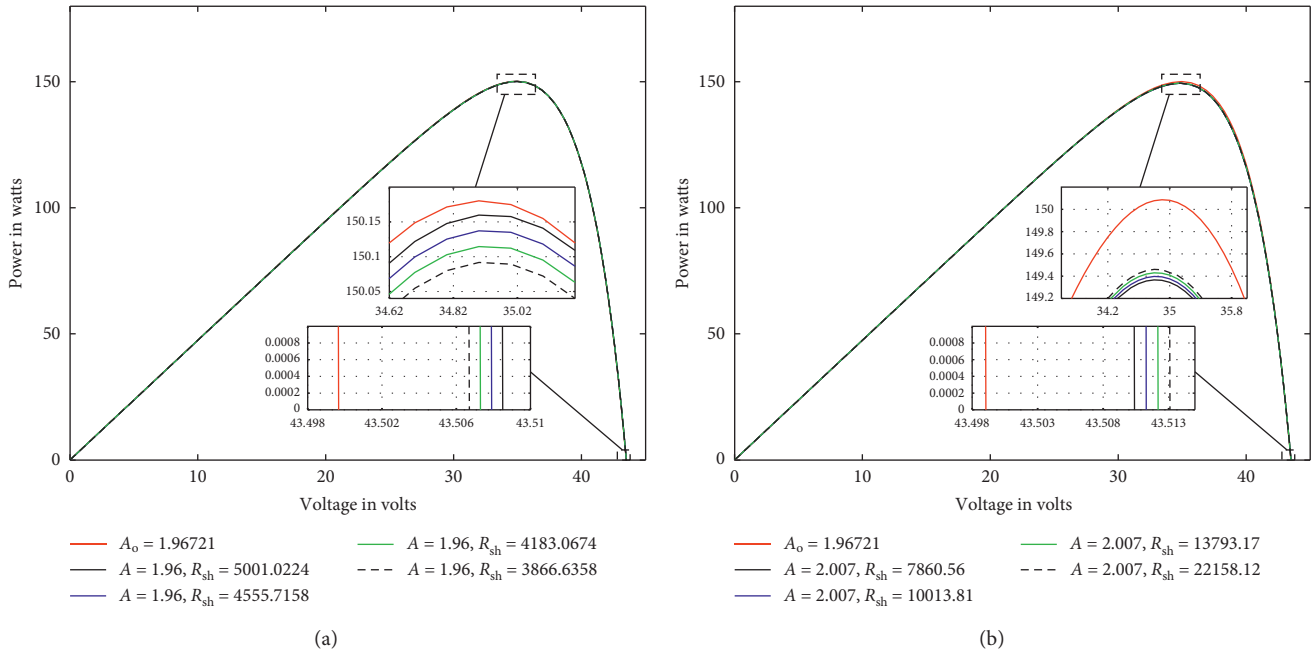


FIGURE 7: PV curves for BP-SX 150 showing different (R_{sh}) values for (a) $A < A_o$ and (b) $A > A_o$.

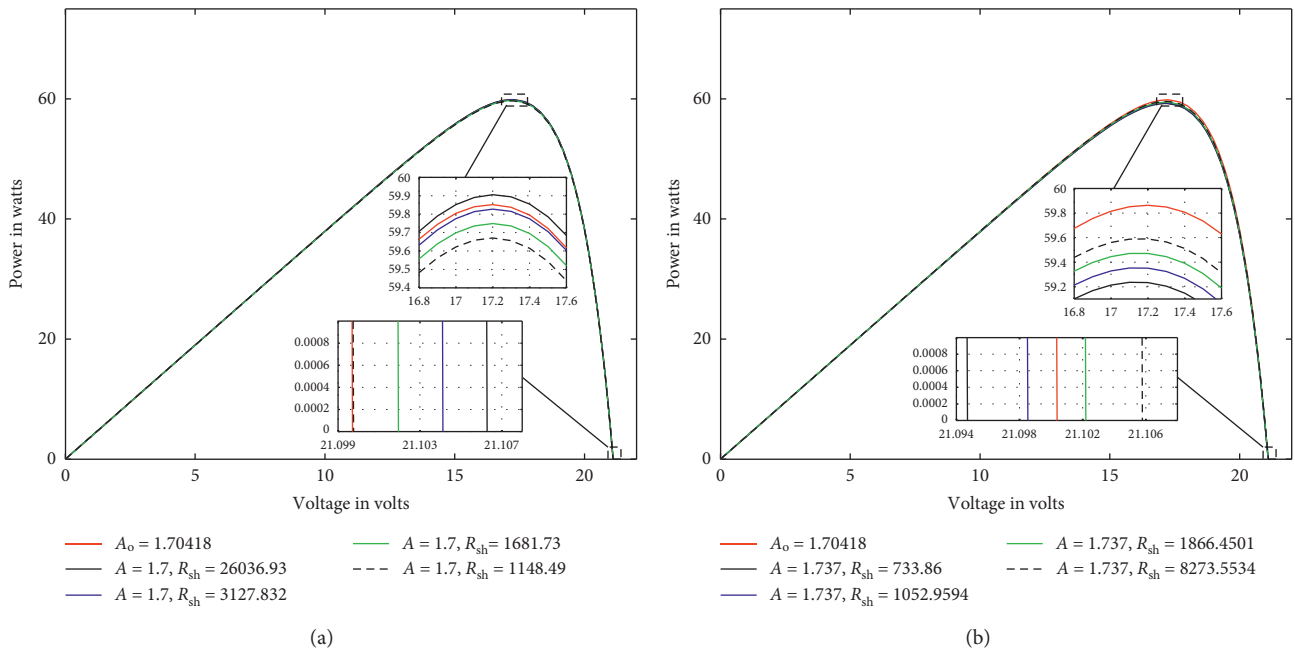


FIGURE 8: PV curves for MSX60 showing different (R_{sh}) values for (a) $A < A_o$ and (b) $A > A_o$.

Depending on which parameters are calculated first, different approaches result to different values of I_o . For instance, if A , R_s , and R_{sh} are calculated first, then equations (13) and (15) can be used to calculate I_o ; otherwise, simplified approaches found in [6, 19] that first assume $R_s \approx 0$ and $R_{sh} \approx \infty$ to obtain I_o have been applied here using equations (14), (16), and (18). Equation (17) gives negative values of I_o and has not been considered here. This accounts for the diverse findings published in previous works such as works

of [25, 71] for BP32235T, findings in [52, 55, 58, 61, 66] for Kyocera 200GT, reports in [61, 72, 73] for BP-SX 150, and studies in [25, 61, 70, 72] for MSX60. The studies in [52, 55, 57–59, 61, 71] used equation (13) to calculate I_o , which provided reliable results that are consistent with the results obtained in this study.

For BP32235T given in Table 7, consistent findings in the works of [25, 71] recorded appropriate values for R_s , R_{sh} , and I_{ph} within the same range as the values obtained in this work.

TABLE 7: Calculated parameters and data from simulations for BP3235T.

Parameters	Method 1	Method 2	Ref. [25]	Ref. [71]
R_s (Ω)	0.2420	0.2420	0.336	0.2371
R_{sh} (Ω)	1066.3747	1066.30	652.09	4393.8
A	1.7920	1.8626	1.01	1.34
I_o (A)	$1.194159E-05$	$1.989991E-5$	$4.610E-10$	$1.29E-07$
I_{ph} (A)	8.481938	8.4819	8.4929	8.4805
I_o (A), equation (3) [55]	$1.19416E-05$	$1.99000E-05$	$3.52066E-10$	$1.26861E-07$
I_{o_s} equation (3)	$1.18925E-05$	$1.98181E-05$	$3.49697E-10$	$1.26735E-07$
I_{o_s} equation (13) [61]	$1.18952E-05$	$1.98227E-05$	$3.49879E-10$	$1.26741E-07$
I_{o_s} equation (14)	$1.19416E-05$	$1.98999E-05$	$3.52066E-10$	$1.26861E-07$
I_{o_s} equation (15)	$5.78361E-06$	$8.93832E-06$	$4.78243E-10$	$1.27022E-07$
I_{o_s} equation (16)	$1.21197E-05$	$1.82458E-05$	$2.84624E-09$	$3.17969E-07$
I_{o_s} equation (18)	$1.19285E-05$	$2.00357E-05$	$3.30414E-10$	$3.17969E-07$

TABLE 8: Calculated parameters and data from simulations for KC200GT.

Parameters	Method 1	Method 2	Ref. [52]	Ref. [55]	Ref. [58]	Ref. [61]	Ref. [66]	Ref. [59]
R_s (Ω)	0.2188	0.2188	0.217	0.221	0.231	0.217	0.2108	0.21095
R_{sh} (Ω)	953.941	679.35	951.927	415.405	594.851	951.92	145.083	192.757
A	1.809	1.882	1.34	1.3	1.3	1.342	1.1578	1.1482
I_o (A)	$1.6535E-5$	$2.7501E-5$	$1.7100E-7$	$9.825E-8$	$9.6990E-8$	$1.71E-7$	$1.01E-8$	$8.6369E-9$
I_{ph} (A)	8.2119	8.2127	8.212	8.214	8.213	8.211	8.226	8.218985
I_o [55]	$1.6535E-5$	$2.7501E-5$	$1.6782E-7$	$9.7328E-8$	$9.7328E-8$	$1.7230E-7$	$1.0340E-8$	$3.665E-9$
I_{o_s} equation (3)	$1.6466E-5$	$2.7339E-5$	$1.6711E-7$	$9.6389E-8$	$9.6672E-8$	$1.7158E-7$	$1.0054E-8$	$3.568E-9$
I_{o_s} equation (13)	$1.6470E-5$	$2.7348E-5$	$1.6715E-7$	$9.6440E-8$	$9.6710E-8$	$1.7162E-7$	$1.0069E-8$	$3.578E-9$
I_{o_s} equation (14)	$1.6535E-5$	$2.7501E-5$	$1.6782E-7$	$9.7327E-8$	$9.7327E-8$	$1.7230E-7$	$1.0340E-8$	$3.665E-9$
I_{o_s} equation (15)	$8.2468E-6$	$1.2462E-5$	$1.6784E-7$	$9.7440E-8$	$9.6720E-8$	$1.7164E-7$	$1.1867E-8$	$1.835E-9$
I_{o_s} equation (16)	$1.6783E-5$	$2.5205E-5$	$4.2778E-7$	$2.7675E-7$	$2.7675E-7$	$4.3690E-7$	$4.6099E-8$	$2.012E-8$
I_{o_s} equation (18)	$1.6516E-5$	$2.7700E-5$	$1.6014E-7$	$9.2594E-8$	$9.2594E-8$	$1.6445E-7$	$9.7438E-9$	$3.443E-9$

TABLE 9: Calculated parameters and data from simulations for BP-SX 150.

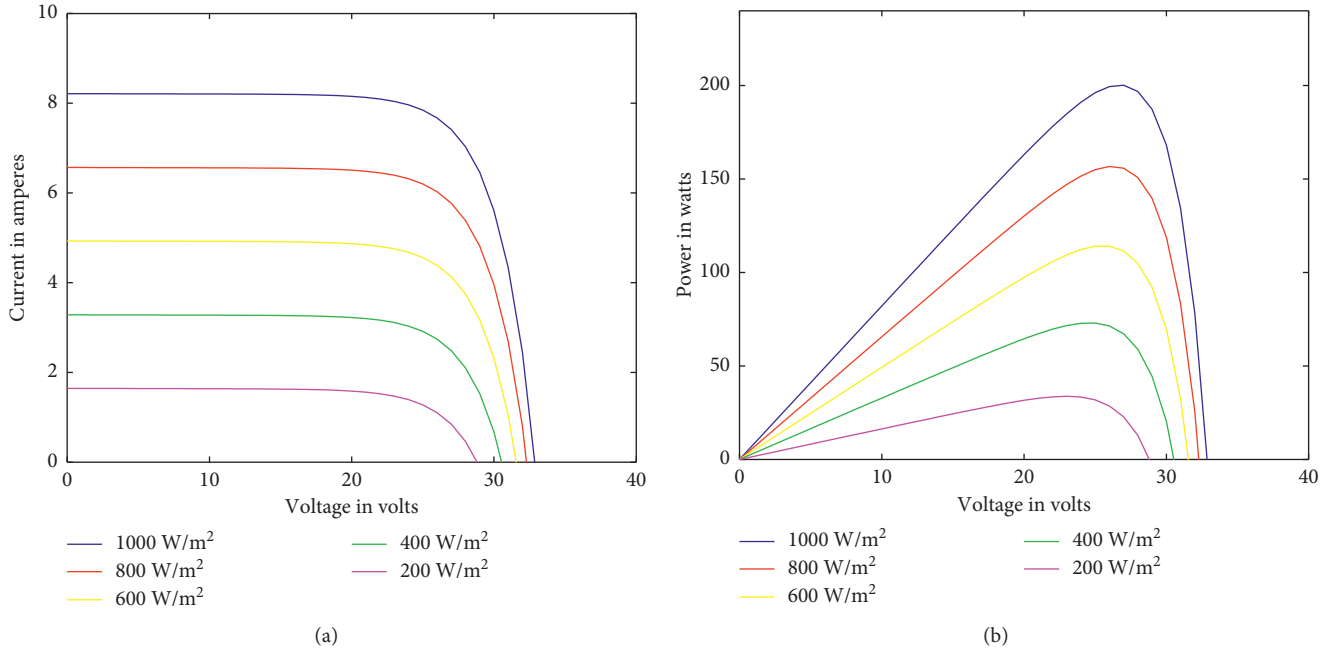
Parameters	Method 1	Method 2	Ref. [61]	Ref ([72])	Ref ([73])
R_s (Ω)	0.3301	0.3301	0.4543	0.312557	0.331466
R_{sh} (Ω)	3866.64	22158.12	960.06	1799.371625	4367.59648
A	1.96	2.007	1.4851	1.64	1.642
I_o (A)	$2.90755E-05$	$3.851289E-05$	$6.166E-07$	$2.8016100E-06$	$2.8419820E-06$
I_{ph} (A)	4.7504	4.7501	4.7522	4.750827	4.75
I_{o_s} equation (3) [55]	$2.907545E-05$	$3.851320E-05$	$6.258569E-07$	$2.794604E-06$	$2.843865E-06$
I_{o_s} equation (3)	$2.900659E-05$	$3.849728E-05$	$6.198869E-07$	$2.780381E-06$	$2.837902E-06$
I_{o_s} equation (13) [61]	$2.900917E-05$	$3.849802E-05$	$6.201831E-07$	$2.780867E-06$	$2.838119E-06$
I_{o_s} equation (14)	$2.907528E-05$	$3.851289E-05$	$6.258568E-07$	$2.794602E-06$	$2.843864E-06$
I_{o_s} equation (15)	$1.930475E-05$	$2.480836E-05$	$6.196016E-07$	$2.784494E-06$	$2.830482E-06$
I_{o_s} equation (16)	$2.934115E-05$	$3.666915E-05$	$1.397445E-06$	$4.578563E-06$	$4.642457E-06$
I_{o_s} equation (18)	$2.905107E-05$	$3.869178E-05$	$5.956163E-07$	$2.697940E-06$	$2.746036E-06$

TABLE 10: Calculated parameters and data from simulations for Solarex MSX60.

Parameters	Method 1	Method 2	Ref. [25]	Ref. [61]	Ref. [70]	Ref. [72]
R_s (Ω)	0.150	0.150	0.234	0.169	0.2165	0.223427
R_{sh} (Ω)	1148.49	8273.5534	9.29	637.5	274.937	406.346152
A	1.7	1.737	1.27	1.404	1.277	1.387
I_o (A)	$5.60956E-6$	$7.46677E-6$	$5.95E-8$	$3.29E-07$	$6.4500E-8$	$2.97156E-6$
I_{ph} (A)	3.800013	3.8001	3.8	3.801	3.813	3.872128
I_{o_s} equation (3) [55]	$5.60957E-06$	$7.46678E-06$	$5.95254E-08$	$3.30851E-07$	$6.56881E-08$	$2.71080E-07$
I_{o_s} equation (3)	$5.58245E-06$	$7.46177E-06$	$2.39471E-08$	$3.27969E-07$	$6.43615E-08$	$2.67376E-07$
I_{o_s} equation (13) [61]	$5.58319E-06$	$7.46191E-06$	$2.54464E-08$	$3.28057E-07$	$6.44132E-08$	$2.67525E-07$
I_{o_s} equation (14)	$5.60956E-06$	$7.46677E-06$	$5.95254E-08$	$3.30851E-07$	$6.56881E-08$	$2.71080E-07$
I_{o_s} equation (15)	$3.84169E-06$	$5.09713E-06$	$-3.60728E-07$	$3.28789E-07$	$6.41383E-08$	$2.26480E-07$
I_{o_s} equation (16)	$5.64471E-06$	$7.11705E-06$	$1.41800E-07$	$5.69360E-07$	$1.53585E-07$	$4.84460E-07$
I_{o_s} equation (18)	$5.60657E-06$	$7.49835E-06$	$5.67053E-08$	$3.19383E-07$	$6.26165E-08$	$2.61219E-07$

TABLE 11: Calculated parameters and data from simulations for KC200GT.

Datasheet values						
I_{sc} (A)	I_{mpp} (A)	V_{oc} (V)	V_{mpp} (V)	K_I (A/°C)	K_V (V/°C)	NOCT (°C)
6.62	6.13	29.9	23.2	0.00328	(-0.123)	47
Simulated parameter values						
Parameters	R_s (Ω)	R_{sh} (Ω)	A	I_o (A)	I_{ph} (A)	T (K)
$T = 25^\circ\text{C}$	0.2188	1028.685	1.809	$1.6535E-05$	6.5695	298.15
$T = 47^\circ\text{C}$	0.233	679.35	1.808	$1.0752E-04$	6.6257	320.15
Ref. [59]	0.23901	218.40213	1.11933	$4.94422E-09$	6.57518	298.15

FIGURE 9: IV and PV curves for KC200GT showing different irradiance values at 1000 W/m^2 , 800 W/m^2 , 600 W/m^2 , 400 W/m^2 , and 200 W/m^2 , for $T = 25^\circ\text{C}$. (a) KC200GT IV curve for $T = 25^\circ\text{C}$. (b) KC200GT PV curve for $T = 25^\circ\text{C}$.

Nevertheless, no consistency exists between A and I_o values, where higher values of ideality factors (A_s) give large I_o values and lower values of ideality factors (A_s) give smaller values of I_o .

Consistent results have also been reported for Kyocera KC200GT as shown in Table 8 where the works of [55, 58] gave closer results for $R_s = 0.221(\Omega)$ and $R_s = 0.231(\Omega)$, while the findings reported in [52, 61, 66] gave $R_s = 0.217(\Omega)$, $R_s = 0.217(\Omega)$, and $R_s = 0.2108(\Omega)$, respectively, which agrees with $R_s = 0.2188(\Omega)$ reported here. The R_{sh} values differ with a higher margin, since a minor adjustment in R_s in $\text{m}\Omega$ s is significant enough to shift R_{sh} exponentially according to equations (32) and (35). The results of [55, 58] are higher than those of methods 1 and 2 and those published in [52, 61, 66] due to the fact that the solutions were obtained for a fixed value of the ideality factor $A = 1.3$. In addition, the different choice of ideality factor gives diverse values of I_o .

For BP-SX 150 data results shown in Table 9, similar and almost matching results were obtained in [73] of $R_s = 0.331466(\Omega)$ compared with the proposed method in this article of $R_s = 0.3301(\Omega)$. The works of [61, 72] reported higher and lower values of $R_s = 0.4543(\Omega)$ and

$R_s = 0.312557(\Omega)$, respectively. Other parameters do not have satisfactory agreements with the parameters extracted here.

For MSX60 data given in Table 10, it is possible to compare the result of $R_s = 0.15(\Omega)$ using methods 1 and 2 with the results given in [61] of $R_s = 0.169(\Omega)$ which shows a slight difference. Nevertheless, the results in [25, 70, 72] have much larger R_s values depending on the techniques applied to achieve them, while diverse results are given for other parameters.

8. Evaluation of the Proposed Method at Different Irradiance Levels and Nominal Operating Cell Temperature

The proposed technique has also been tested for various irradiance levels and nominal cell operation temperature (NOCT). Table 11 gives a summary of Kyocera KC200GT nominal operation cell temperature data that have been provided in manufacturer's datasheet. Figures 9 and 10 show

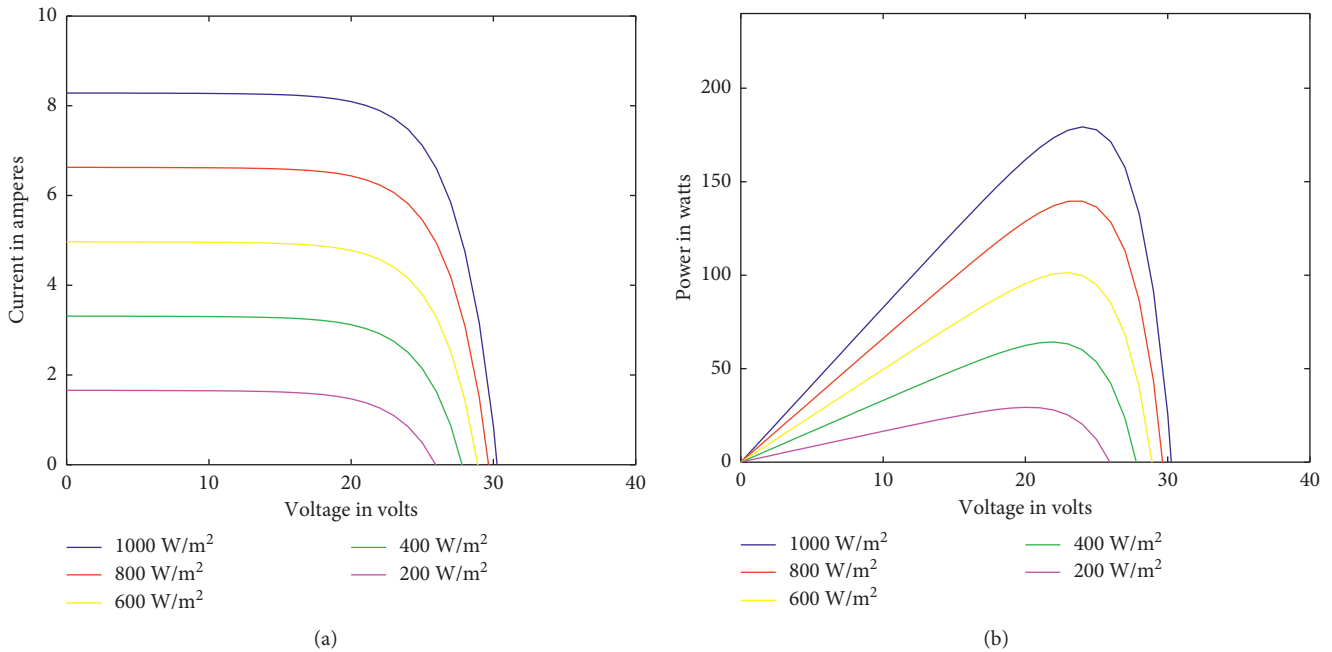


FIGURE 10: IV and PV curves for KC200GT showing different irradiance values at 1000 W/m^2 , 800 W/m^2 , 600 W/m^2 , 400 W/m^2 , and 200 W/m^2 , for $\text{NOCT} = 47^\circ\text{C}$. (a) KC200GT IV curve for $\text{NOCT} = 47^\circ\text{C}$. (b) KC200GT PV curve for $\text{NOCT} = 47^\circ\text{C}$.

the IV and PV curves at both STC and NOCT for KC200GT. It is clear from the figures that the IV and PV characteristics coincide at a satisfactory level with the experimental data provided in the datasheet.

9. Conclusion

A new method has been developed and tested to determine the unknown parameters of a single-diode model for PV modules based on the datasheet values. This new algorithm provides a simplified and robust way of extracting A , I_o , R_{sh} , R_s , and I_{ph} from the transcendental single-diode PV model equation using IV curve crucial points from the manufacturer's data. Further, these parameters have been used to plot the PV characteristics for the solar modules whose three critical points, I_{sc} , P_{mpp} , and V_{oc} match the datasheet values or data obtained through an experimental procedure. Various values of ideality factors near the optimum ideality factor offer different options of selecting the series and shunt resistance pairs that match the experimental data.

Data Availability

The simulation data used to support the findings of this study are included within the article and are available from the corresponding author upon request via ndegwarg@uonbi.ac.ke.

Conflicts of Interest

The authors declare that they have no conflicts of interest.

Acknowledgments

The authors would like to thank the Department of Physics and the Board of Graduate School, Nairobi University (Kenya), for providing all the facilities and financial support, respectively, for this research.

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